

From (6) we can find the optimum  $\theta_0$  spacing value minimizing the insertion loss

$$\theta_0 = \tan^{-1} \frac{2R_0(1 - \omega_0^2 LC)}{\omega_0 [L(2 - \omega_0^2 LC) + CR_0^2]}. \quad (7)$$

If (7) is applied, at  $f = f_0$  a zero of attenuation occurs. A wide band of operation can therefore be used across  $f_0$ , if the elementary low-pass cell characteristics are suitably chosen. In fact, if the tee cell is individually well matched, the same will be true for the overall network. Thus, optimizing the initial tee low-pass filter in order to limit its maximum mismatching loss to 0.20–0.25 dB ( $X \leq 1.15$ ), and choosing  $f_0$  about coinciding with the frequency corresponding to maximum of ripple, i.e., to maximum mismatching (this frequency, see (5), is  $f_z/\sqrt{3}$ ), the best performance can be obtained in a multi octave band (see Fig. 4). Fig. 5 shows the

return loss of a double  $\theta_0$ -spaced tee cell configuration, in the case  $X = 0.9$ . From (5), maximum mismatching occurs at  $f = 10$  GHz. Choosing  $f_0 = 10$  GHz, applying (7) yields  $\theta_0 = 45^\circ$  at 10 GHz. As we can see from the figure, when  $\theta_0 = 45^\circ$  the network is well matched at 10 GHz, and in the range 2–18 GHz, VSWR is less than 1.35:1. When  $\theta_0$  is different from  $45^\circ$ , VSWR increases. However, for  $30^\circ \leq \theta_0 \leq 60^\circ$ , we note a maximum VSWR of about 1.5:1. Outside this electrical length range we observe a relevant degradation of performance. Isolation typically results 40–50 dB. Higher values can be obtained increasing the number of  $\theta$ -spaced tee cells.

## REFERENCES

- [1] L. Altman, *Microwave Circuits*, New York: Van Nostrand, 1964.
- [2] "Reducing the insertion loss of a shunt PIN diode," HP Application Note #957-2.
- [3] J. F. White, *Semiconductor Control*, Dedham, MA: Artech House, 1977.

## Letters

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### Correction to "Asymptotic High-Frequency Modes of Homogeneous Waveguide Structures with Impedance Boundaries"

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The treatment of Section V in the above paper<sup>1</sup> was incomplete and, as such, a bit misleading. In fact, the existence question for solutions of equation (53) for  $f_2 = (\pi_2, m_2)$  did not properly take into account the degeneracy of the basic modes  $f_1 = (\pi_1, m_1)$ . It is known that for a solution to exist, the right-hand side of a deterministic equation like (53) must be orthogonal to all solutions of the homogeneous adjoint problem, which in this case is the basic problem with solutions  $f_1$ . Without degeneracy, equation (56) would be that condition. However, since there are at least two linearly independent solutions  $f_{1i}$ , there are at least two such conditions, which leads to a contradiction except if  $f_1$  in (53) is chosen in a special way. Let us denote the admissible  $f_1$  in (53) by  $f'_1$  and it can be written as a linear combination of any complete set of degenerate basic modes corresponding to the same parameter  $\beta_1$ :  $f'_1 = \sum \alpha_i f_{1i}$ . The orthogonality condition reads

$$(f'_1, Lf_2) - (f'_1, Bf_2)_b = 2\beta_2(f'_1, f'_1) - 2j\beta_1(f'_1, Mf'_1)_b = 0 \quad (1)$$

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<sup>1</sup>I. V. Lindell, *IEEE Trans. Microwave Theory Tech.*, vol. MTT-29, pp. 1087–1091, Oct. 1981.

and it should be satisfied for all  $i = 1 \dots n$  ( $n \geq 2$ ). This is an algebraic equation for the matrix  $\alpha = (\alpha_i)$

$$j\beta_1 M \cdot \alpha = \beta_2 F \cdot \alpha \quad (2)$$

where we denote  $M = (f_{1i}^*, Mf_{1j})$ ,  $F = (f_{1i}^*, f_{1j})$ . The unknown coefficient  $\beta_2$  is obtained as a solution of the eigenvalue equation

$$\det(j\beta_1 M - \beta_2 F) = 0 \quad (3)$$

and the corresponding admissible right-hand side of a (53) from the eigenvectors  $\alpha$  of (2).

The expression (56) giving  $\beta_2$  in terms of  $f_1$  is not incorrect but ambiguous, because if we would substitute  $f'_1$  for  $f_1$  we would obtain the correct  $\beta_2$ . Hence, the general conclusions following (56) in Section V are valid if  $f'_1$  is understood everywhere in place of the ambiguous  $f_1$ .

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## REFERENCES

- [1] C. Dragone, "High-Frequency behavior of waveguides with finite surface impedances," *Bell Syst. Tech. J.*, vol. 60, no. 1, Jan. 1981, pp. 89–116.